

# Chromatic Dispersion of Thin Film Filters

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**Abstract:** This paper discusses the chromatic dispersion effects of thin film filters used in telecommunications systems. ITU channel, banded architecture, and low dispersion filter results are presented. Design techniques to improve and manage chromatic dispersion in practical network implementations are discussed.

## 1. Introduction

Optical thin film filters (TFFs) have been widely deployed in telecommunications optical networks for a number of applications including channel multiplexing and demultiplexing, optical add drop, banded architecture multiplexing and demultiplexing, and dichroic bandsplitting among others. The deployment of thin film filters has been driven by their highly desirable attributes such as ease of manufacture, thermal stability, and flat, square filter spectral profiles. In most cases the chromatic dispersion (CD) inherent in TFFs is not a significant concern; however, as bit rates and optical channel densities continue to increase, the issue of CD in TFFs has grown in importance. This paper discusses chromatic dispersion effects of TFFs both in theory and as implemented in modern networks.

## 2. Thin film filter background

Thin film filters (TFFs) are typically designed as a set of stacked Fabry-Perot cavities with the mirrors consisting of a stack of high and low index quarterwave dielectric layers [1]. The reflectivity of the mirrors of each cavity, the properties of the spacers between the mirrors, and the number of cavities determine the optical properties such as the bandwidth and spectral slopes. With narrowband ITU channel filters, which are typically used to multiplex (mux) or demultiplex (demux) individual channels, the group delay (GD) peaks near the filter 3dB points and gives a characteristic group delay as shown in Fig. 1 with a shape resembling “bat ears”. For this filter type, CD grows rapidly as channel density increases. For a 50 GHz ITU channel drop filter, group delay variation from peak to valley is twice as large as that of a 100 GHz filter of similar design. Further, the 50 GHz filter transmission band is half as wide as that of the 100 GHz filter, with the group delay peak half as far from the filter center. Thus the maximum value of CD, the derivative of group delay with respect to wavelength, is approximately 4 times higher for the 50GHz filter compared to the 100 GHz filter (with identical adjacent channel isolation). Similarly, the maximum value of CD for a 25 GHz filter is roughly 16 times that of the 100GHz filter [2].

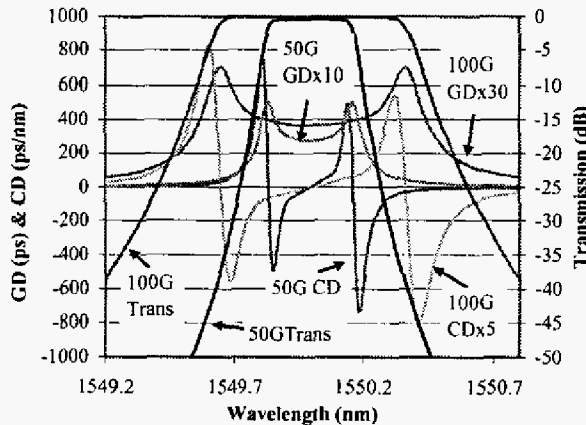


Fig. 1. 50GHz and 100GHz GD and CD plots.

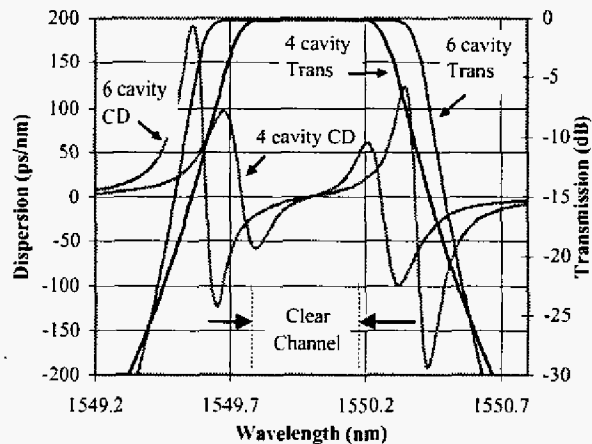


Fig. 2. Effect of cavity count on CD for 100GHz flat top

The optical performance of a thin film filter can be mathematically described as an infinite impulse response (IIR) filter [3]. In transmission a TFF is also a minimum phase filter and thus the transmission amplitude response of the filter is directly related to the phase response via a Hilbert transform or the equivalent Kramers-Kronig

relationship. It follows that the CD of a bandpass TFF is unambiguously determined from the transmission properties. Another outcome of the relationship is a useful “rule of thumb” that the peak value of group delay is directly proportional to the rate of change of the slope of the transmission.

Since the relationship between phase and amplitude is a Hilbert transform, it is possible improve the CD within the passband while minimizing the effect on the amplitude over that same region by making the changes outside of the central passband region. Increasing the cavity count of the filter design allows the 3dB bandwidth to increase while maintaining transmission isolation, reducing inband CD as shown in Fig. 2. Another approach to reducing CD is to slightly modify the transmission profile as shown in Fig. 3. The flatness of the group delay results from the rounded transmission profile in the center of the filter [4,5]. If the flatness of the amplitude response of a bandpass TFF is not critical, then the sharp spectral band edges of the filter can be reduced to the point of completely eliminating the group delay “bat ears” and the resultant filter has flat group delay, with near zero CD as shown in Fig. 4. Filter rounding can be even further increased beyond this point of minimal CD to create a filter with CD slope opposite to that of a flat top design.

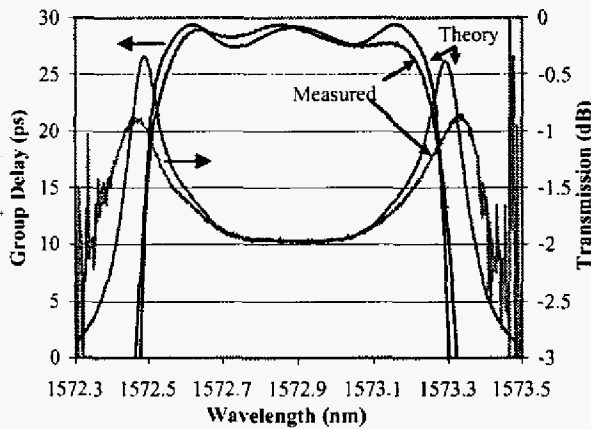


Fig. 3. 100GHz filter designed for low dispersion using a rounded amplitude response.

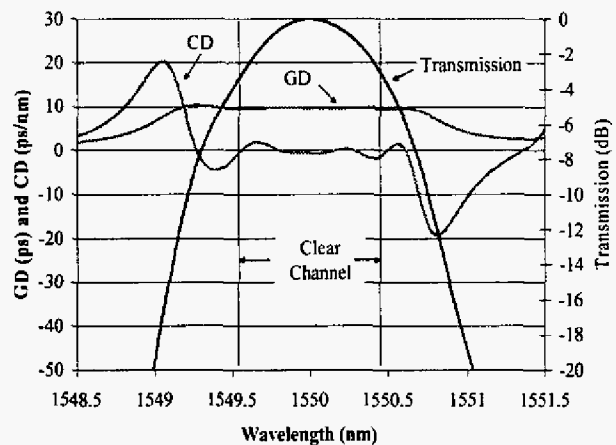


Fig. 4. 100GHz rounded top filter design with minimal CD.

It is worth noting that the peak value of CD of a TFF over the passband cannot be compared directly to the CD value of optical fiber because the filter CD is not constant in wavelength. The group delay is symmetric about the filter center wavelength and so negative and positive sidebands of the optical carrier centered on the filter see the same (positive valued) group delay. In a length of fiber, these sidebands see opposite signed group delay relative to the carrier. The latter causes more distortion than the former. The result is that the dispersion power penalty of a narrow bandpass TFF is less than that of an equivalent fiber of the same peak CD value. Accurate analysis requires numerical simulations to obtain a prediction of its effect in a given transmission system.

The previous discussion has focused on the transmission properties of a TFF. With symmetric TFF designs (i.e. where the design is the same when reversing the order of the layers) the reflection is also minimum phase and the Hilbert transform relationship applies to the reflected light as well. If the filter is asymmetric then the relationship between phase and amplitude is not restricted by the Hilbert transform and the designer has, in general, much more freedom to change the CD relative to the amplitude. This is demonstrated in particular with Gires-Tournois (GT) interferometer thin film devices such as is discussed in section 6.

### 3. Banded architecture (Skip filters)

TFF “skip” filters split, mux/demux, or add and drop continuous bands of channels. They are popular for their “pay-as-you-grow” nature, providing a simple and inexpensive method of increasing capacity by adding groups of channels to previously installed links. Fig. 5 shows a measured example of a 10 skip 0 filter for 100 GHz channel spacing [6]. The “10 skip 0” designation means that 10 channels can pass in transmission through the filter and all other ITU channels are completely reflected. CD of these types of filters is identical to those discussed above except that the channels are now a small fraction of the filter width. Channels can reside on either the edge of the filter where the CD is largest or at the center where the CD is zero. Channels on opposite edges have opposite CD slopes.

In contrast, a 4 skip 1 filter would pass 4 channels in transmission but the channel at the filter skirt is partially transmitted and partially reflected causing the channel to be lost (or "skipped").

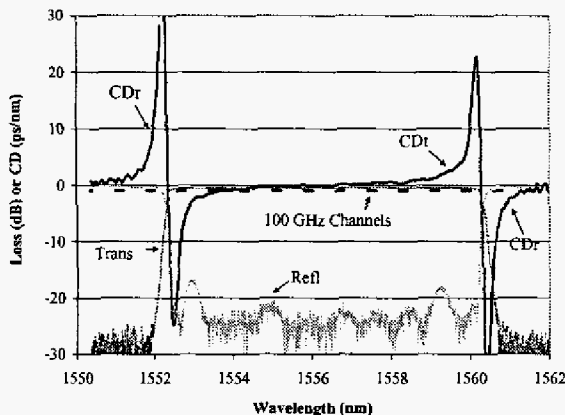


Fig. 5. Measured CD in Reflection (CDr) and Transmission (CDt) of 10s0

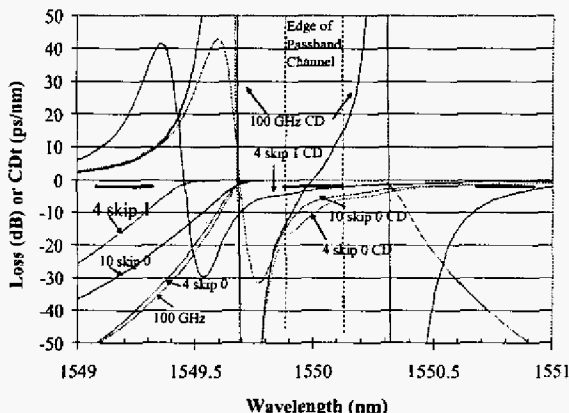


Fig. 6. Calculated loss and CDt of various banded filters.

For practical applications, how does the CD of a skip filter compare to that of a narrowband filter? Fig. 5 shows a comparison of 100GHz versions of a 10 skip 0, 4 skip 0, and 4 skip 1 as well as a 100 GHz single channel filter with all of the filters lined up at a channel at the edge of the passband. In this figure, the 100 GHz channel filter and the 4 skip 0 filter are designed for 25 dB adjacent channel isolation and flat top passbands. The CD in transmission for both the 100 GHz channel filter and the 4 skip 0 are very similar at the edge of the filter passbands. The steepness of the 4 skip 1 and the 10 skip 0 functions are similar leading to similar shapes of the CD. The 4 skip 1 filter however has a lower effective CD for its outer channel, because the outer channel is farther from the filter edge due to the skipped channel.

#### 4 Packaged filter considerations

The optical performance of a packaged TFF is affected by a number of factors including beam divergence, thermal stability, wavelength centering, among others. Wavelength centering is particularly important as error in centering the filter during the packaging process increases CD since the CD is lowest at the center wavelength.

Figure 7 shows the CD of a TFF before and after packaging. The transmission and reflection spectral profiles are somewhat rounded in the packaged filter case. The passband is narrower and the peak CD is lower after packaging due to filter rounding caused by beam divergence as well as other packaging issues.

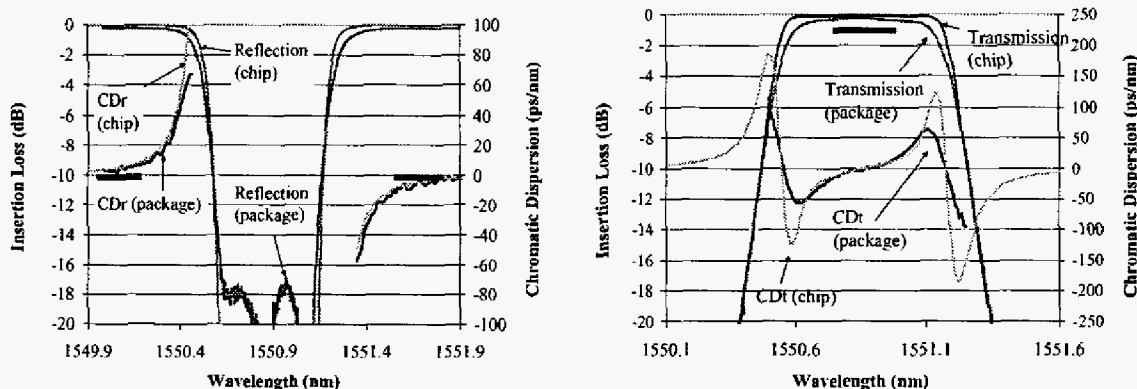


Fig. 7. Amplitude and CD of TFF filter before (chip) and after packaging for reflection in chart on left and transmission on the right.

#### 5. Module Considerations

Modules are constructed by cascading TFFs in a three port package via splicing. A three port package aligns the input/common, reflection and transmission signals to individual optical fibers. In a module cascade, the worst case channel CD is the sum of the transmission CD and the adjacent channel reflection CD. Figure 8 show two module

architectures, the first being a simple cascade and the second using a 3dB coupler to create two legs of interleaved channels. The interleaved channels reduce total CD by increasing the adjacent channel spacing (which reduces reflection CD). The 3dB coupler trades off increased loss for lower CD in this design.

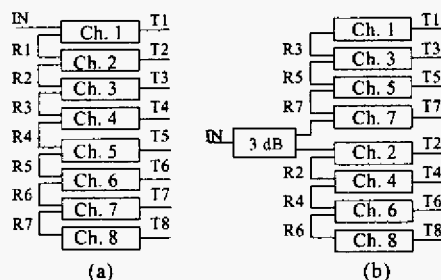


Fig. 8. Module architectures for TFF. 8a is a simple cascade. 8b trades off loss for lower CD.

## 6. CD compensation using a Gires-Tournois (GT) etalon

The characteristic CD of a flat top TFF can be compensated by a GT etalon designed to have opposite slope and equal magnitude of CD to the TFF filter. An example of this approach is shown in Fig. 9, where a GT compensator filter was fabricated using a single cavity TFF and spliced in series with the transmission signal of a 50 GHz TFF filter. As can be seen in the figure, the CD has been reduced across the passband, and nearly eliminated in the center 100 picometers.

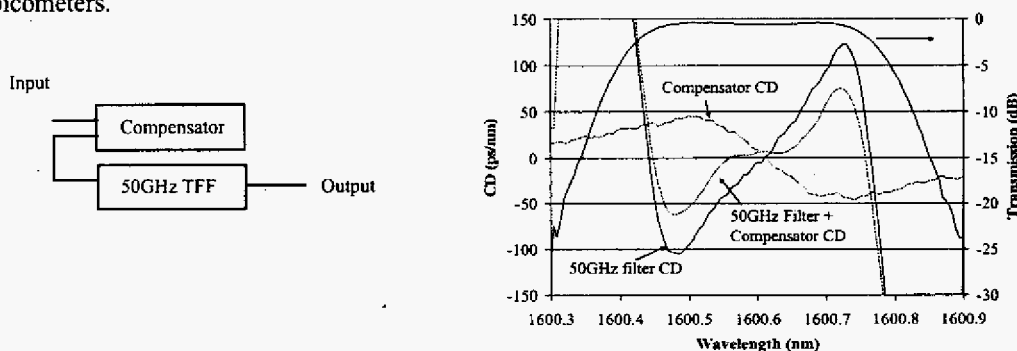


Fig. 9. Chromatic dispersion of 50GHz filter, GT compensator, and combination of both as shown in schematic on left.

## 7. Summary

Thin film filters have a wide variety of applications in optical communications due to their outstanding spectral performance and low cost. TFFs have inherent chromatic dispersion in both transmission and reflection that can be managed in the design process. Design techniques to lower CD include adjustment of cavity count, varying the passband amplitude shape, and cascading with compensation elements. CD of TFF is also unique relative to fiber as it is zero at the center of the passband, which provides lower inherent dispersion power penalties for a given CD value. CD of TFFs at a module level can be managed by the use of module architecture designs which tradeoff insertion loss for lower CD.

## 8. Acknowledgements

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## 9. References

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